Poles and Zeros Response of Geophones Bill Menke, January 2017

The frequency-dependence of the geophone response is expressed as a Padé Approximant, that is as a ratio of two polynomials, the zeros polynomial $Z(\omega)$ (in the numerator) and poles polynomial $P(\omega)$ (in the denominator):

$$R(\omega) = F \frac{Z(\omega)}{P(\omega)} \text{ with } F = S_r S_g A_0 \text{ and } Z(\omega) = \prod_{n=1}^N (i\omega - z_n) \text{ and } P(\omega)$$
$$= \prod_{m=1}^M (i\omega - p_m)$$

Here ω is angular frequency, measured in radian/s. The zeros z_n and poles p_m are complex constants. Because of the requirement that that time-domain response be purely real, poles and zeroes that lie off the real ω -axis must occur in complex conjugate pairs.

The normalization constant A_0 is defined such that $|A_0| | P(\omega_r)/Z(\omega_r)| = 1$ at the reference frequency ω_r .

The sensitivity constant S_g of the geophone is usually expressed with respect to ground velocity, e.g. in volts per meter/second. The sensitivity of the geophone is therefore exactly S_g at the reference frequency ω_r , since $|A_0| |P/Z| = 1$ at that frequency. To convert to a sensitivity with respect to displacement, consider a harmonic wave with velocity $v(t) = \exp(i\omega_r t)$, which has displacement $v(t) = \exp(i\omega_r t)/(i\omega_r)$. Thus the sensitivity for displacement, say S'_g , expressed in volts per meter, has an added factor of ω_r : $S'_g = \omega_r S_g$.

In this treatment, we assume that the recording process is frequency-independent, so that the conversion from volts to digital counts can be expresses by a sensitivity constant S_r measured in digital counts per volt.

Thus, the output of the recorder, d(t), in digital counts, is related to the ground velocity v(t), in m/s by:

$$d(t) = \mathcal{F}^{-1} \{ R(\omega) \mathcal{F} \{ v(t) \} \}$$

Here \mathcal{F} is the Fourier transform and \mathcal{F}^{-1} is its inverse.

Sometimes, one wants to know how the output of the recorder, d(t), in digital counts, is related to the ground displacement u(t). This requires modifying the zero polynomial by adding a factor $(i\omega - z_{N+1})$ with $z_{N+1} = 0$, since a derivative in the time domain is equal to multiplication by $i\omega$ in the frequency domain:

$$Z'^{(\omega)} = \prod_{n=1}^{N+1} (i\omega - z_n)$$
 with $z_{N+1} = 0$

The new normalization constant, say A'_0 must have an added factor of ω_r in its denominator so that P(z) has unit modulus at the reference frequency. Thus, $A'_0 = A_0/\omega_r$. Note, however, that $S'_g A'_0 = S_g A_0$; thus in practice no explicit modification of the geophone sensitivity and normalization factor is required. The displacement response is therefore:

$$R'(\omega) = F \frac{Z'(\omega)}{P(\omega)}$$

Examples:



Response of the geophone/recorder system to a unit spike in displacement, for the BHE channels of seismic stations TA/J59E (black) and PO/LATQ (red). The time series are of length N = 1024 and have sampling interval $\Delta t = 0.025 \ s$. The displacement time series is a unit spike, that is, it is everywhere zero, except for the value of 1×10^{-6} m at sample N/2.

An instrument response can be deconvolved from the recorded data d(t) to produce an estimate of ground displacement u(t):

$$u(t) = \mathcal{F}^{-1}\left\{\frac{\mathcal{F}\{d(t)\}}{R(\omega)}\right\}$$

However, because $Z'(\omega)$ has at least one, and usually several, zeroes on the real axis (typically at $\omega = 0$), this process is unstable and leads to amplification of low-frequency noise. Array equalization, in which the response of several stations is adjusted to a reference response $R_r(\omega)$, is often a better option. In this case, one works in corrected digital counts, $d_c(t)$:

$$d_{c}(t) = \mathcal{F}^{-1}\left\{ \frac{R_{r}(\omega)}{R(\omega)} \mathcal{F}\{d(t)\} \right\} = \left\{ \frac{Z_{r}(\omega)P(\omega)}{Z'(\omega)P_{r}(\omega)} \mathcal{F}\{d(t)\} \right\}$$

The calculation can be made stable by analytically cancelling common factors in the numerator and denominator of $Z_r(\omega)/Z'(\omega)$. For example, both numerator and denominator contain the factor $(i\omega - 0)$, and these common factors cancel. (Common factors can also be cancelled from $P(\omega)/P_r(\omega)$, however these polynomials seldom have poles on the real axis, so cancellation is usually unnecessary).



Corrected data $d_c(t)$ for the BHE channels of seismic stations TA/J59E (black) and PO/LATQ (red), array-equalized to the response of TA/J59E. Note that the two time series exactly overlay.